

# Study the Performance of SLM for Different Number of Subcarriers

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**Abstract**— Orthogonal Frequency Division Multiplexing (OFDM) is an attractive modulation technique for transmitting large amounts of digital data over radio waves. One major disadvantage of OFDM is that the time domain OFDM signal which is a sum of several sinusoids leads to high peak to average power ratio (PAPR) which leads to power inefficiency in RF section of the transmitter and increased complexity in the analog to digital and digital to analog Converter. Selected mapping (SLM) is a well-known method for reducing the PAPR in OFDM. In this paper, we have studied the performance of SLM for Different Number of Subcarriers. Simulation result shows that the PAPR is reduced significantly when the number of phase sequences is increased and PAPR is increased when the number of subcarriers is increased. It also shows that data speed increases when subcarriers increase where N-point IFFT/FFT circuit depends on N-subcarriers.

**Index Terms**— OFDM, PAPR, Power Amplifiers, Selected Mapping, Complimentary Cumulative Distribution Function

## I. Introduction

Orthogonal frequency division multiplexing (OFDM) is one of the most attractive techniques for fourth generation (4G) wireless communication [3]. It effectively combats the multipath fading channel, improves the bandwidth efficiency and increases system capacity so as to provide a reliable transmission [2]. OFDM is to split a high rate data stream into a number of lower rate data streams and modulated by subcarriers. These subcarriers are overlapped with each other. The symbol duration

increases for lower rate data streams, the relative amount of dispersion in time caused by multipath delay spread is decreased. Inter symbol interference (ISI) is eliminated almost completely by introducing a guard time in every OFDM symbol [2][3].

OFDM faces several challenges. The key challenges are large peak to average power ratio (PAPR) due to non-linearity of power amplifier, phase noise problems of local oscillator, frequency offset due to Doppler shift or difference between transmitter and receiver. Large peak to average power (PAP) ratio distorts the signal if the transmitter contains nonlinear components such as power amplifiers (PAs). The nonlinear effects on the transmitted OFDM symbols are spectral spreading, inter modulation and changing the signal constellation. The measurement of a waveform, calculated from the peak power of the waveform divided by the average power of the waveform is called crest factor or peak-to-average ratio (PAR) or PAPR.

The PAPR of the OFDM symbol is defined as the ratio of the peak power and the average power [8]:

$$PAPR = \frac{P_{peak}}{P_{average}} = 10 \log_{10} \frac{\max[|x_n|^2]}{E[|x_n|^2]}$$

where  $P_{peak}$  represents output peak power,  $P_{average}$  means output average power,  $[\ ]$  denotes the expected value and  $x_n$  represents the transmitted OFDM signals which are obtained by taking IFFT operation on modulated input symbols [9].

The Crest Factor (CF) is defined as the ratio between maximum amplitude of OFDM signal  $s(t)$  and root mean square (RMS) of the waveform. The CF

is defined as [2]:

$$CF(s(t)) = \frac{\max[|s(t)|]}{E[|s(t)|^2]} = \sqrt{PAPR}$$

Therefore the PAs requires a back off which is approximately equal to the PAPR for distortion less transmission. This decreases the efficiency for amplifiers. Several techniques have been proposed in the literature to reduce the PAPR [1]. These techniques can mainly be categorized in to signal scrambling techniques and signal distortion techniques. Signal scrambling techniques are all variations on how to scramble the codes to decrease the PAPR. Coding techniques can be used for signal scrambling. Golay complementary sequences, Shapiro-Rudin sequences, M sequences, Barker codes can be used efficiently to reduce the PAPR. However with the increase in the number of carriers the overhead associated with

exhaustive search of the best code would increase exponentially. More practical solutions of the signal scrambling techniques are block coding, Selective Level Mapping (SLM) and Partial Transmit Sequences (PTS) [1] [4]. Signal scrambling techniques with side information reduces the effective throughput since they introduce redundancy. The signal distortion techniques introduce both in band and Out-of-band interference and complexity to the system [9]. The signal distortion techniques reduce high peaks directly by distorting the signal prior to amplification. Clipping the OFDM signal before amplification is a simple method to limit PAPR. However clipping may cause large out-of band (OOB) and in-band interference, which results in the system performance degradation. More practical solutions are peak windowing, peak cancellation, Peak power suppression, weighted multicarrier transmission, companding etc.

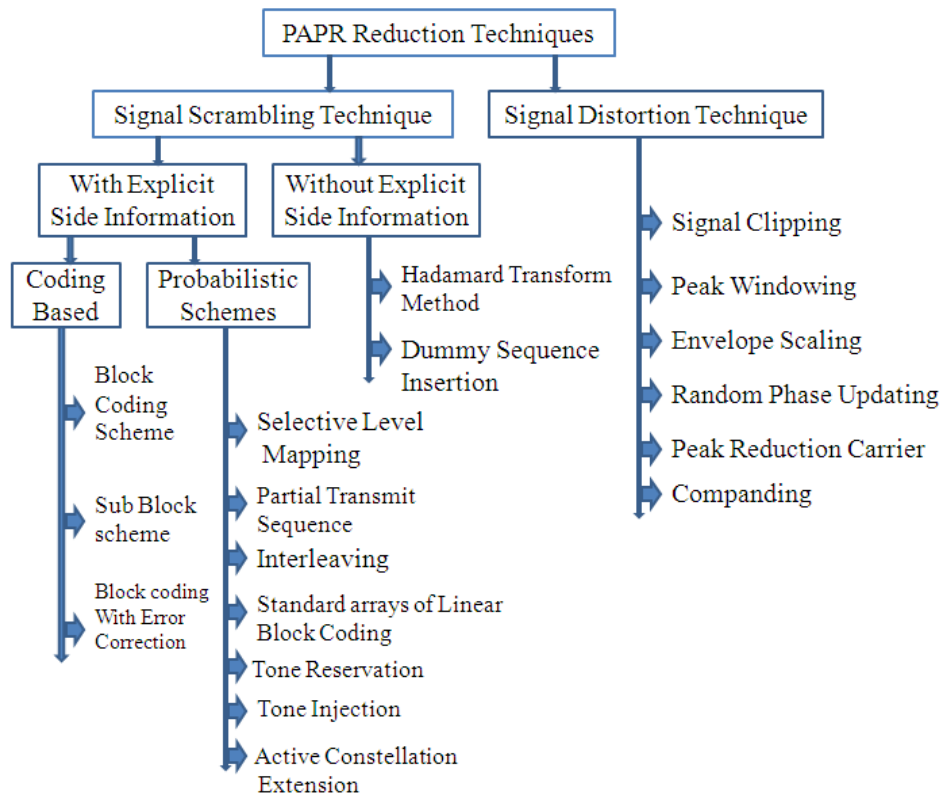


Fig.1: PAPR Reduction Techniques

Basic requirement of practical PAPR reduction techniques include the compatibility with the family of existing modulation schemes, high spectral efficiency and low complexity. Fig. 1 shows the PAPR reduction techniques.

Signal Scrambling has the advantage that it introduces no distortion in the transmitted signal [1] and achieve significant PAPR reduction. It suffers from data rate loss due to transmissions of several side information and need powerful channel code to protect side information. It makes the system more complex

and increases the transmission delay. Other hand Signal Distortion techniques are simple, do not required extra side information. It has following disadvantages: Distortion falls in both in band and out of band. Bit Error Rate increases with increase in number of subcarriers. Out of band radiation reduces spectral efficiency [10]. In this paper, we have studied the performance of SLM for Different Number of Subcarriers in OFDM system.

The remainder of this paper is organized as follows: Section 2 OFDM technology is introduced and a basic

system model of OFDM is presented. Section 3 describes the selected mapping technique (SLM) and complementary cumulative distribution function (CCDF). The simulation result and discussion are given in section 4. And finally, conclusion is drawn in Section 5.

### II. OFDM System Model

Fig. 2 shows the system model considered in this paper. At the starting end binary data is being inputted to the system. After that this binary data is going through to the process of digital to analog mapping. And this mapped signal is being modulated with proper modulation technique. This modulation signal is then converted into parallel signal by serial to parallel converter. This parallel data is being inputted to the IFFT operation block.

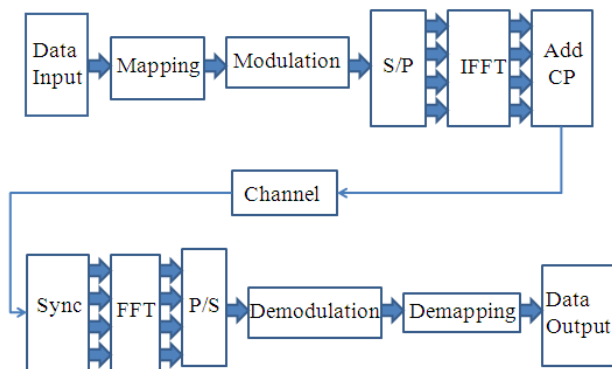


Fig. 2: OFDM transceiver block diagram

This IFFT block converts the frequency domain signal to time domain signal for each subcarrier [7]. Cyclic prefix is being added with the time domain signal. After that this time domain signal is being transmitted through the channel after proper filtering and applications [8]. This transmitted signal is received by receiver and the receiver action is somewhat inverse

of transmitter [5]. The starting end of receiver is a synchronizer. It is being used to equalize the received signal to prevent any channel effect and to get the proper signal.

### III. Selected Mapping (SLM) Method

The complementary cumulative distribution function (CCDF) of the original signal sequence PAPR above a threshold  $PAPR_0$  is written as  $Pr(PAPR > PAPR_0)$ . Accordingly K statistical independent signal waveforms, CCDF can be rewritten as  $[\{Pr(PAPR > PAPR_0)\}]^k$  so that the probability of PAPR that exceeds the same threshold will drop to a small value [3].

The probability of PAPR larger than a threshold  $z$  can be written as  $P(PAPR > z) = 1 - (1 - \exp(-z))^N$ . In the discrete time domain, an OFDM signal  $x_n$  of  $N$  subcarriers can be expressed as [6][8]

$$x_n = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} X_k e^{j2\pi kn/N}, \quad 0 \leq n \leq N-1 \quad (1)$$

Where,  $k = 0, 1, 2, 3, \dots, N-1$  are input symbols modulated by BPSK or QAM and  $n$  is the discrete time index. As to the discrete time signals, since symbol-spaced sampling may sometimes miss some of the signal peaks, signal samples are obtained by oversampling by a factor of  $L$  to better approximate the true  $PAPR$ . Oversampled time-domain samples are usually obtained by  $LN$ -point IFFT of the data block with  $(L-1)N$  zero padding.  $L=4$  is sufficient to capture the peaks. In the SLM approach,  $U$  statically-independent phase sequence. Fig 3 shows the block diagram of Selective Mapping (SLM) technique for PAPR reduction [6].

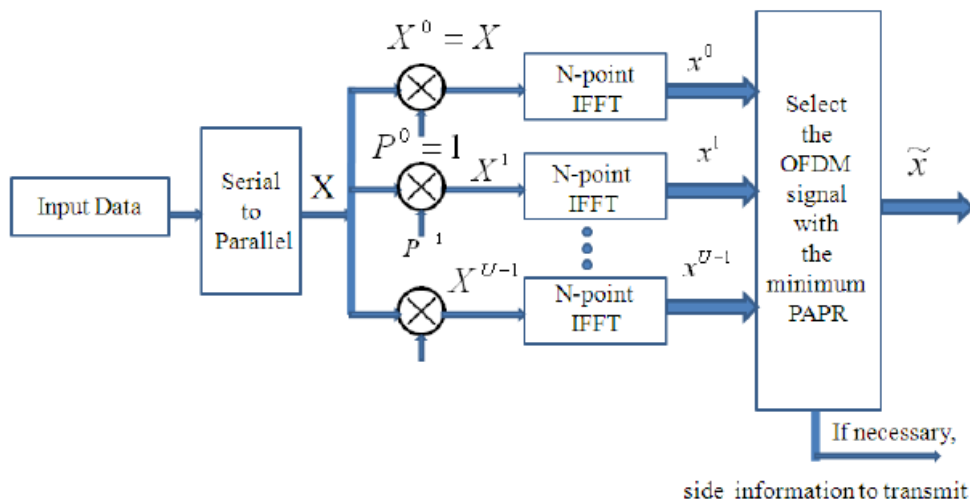


Fig. 3: Block diagram of Selective Mapping (SLM) technique

Here, the input data block  $X = [X|0, X|1, \dots, X|N-1]$  is multiplied with  $U$  different phase sequences  $P^u = [P^u_0, P^u_1, \dots, P^u_{N-1}]^T$  where  $P^u_v = e^{j\phi^u}$  and  $\phi^u_v \in [0, 2\pi]$  for  $v = 0, 1, \dots, N-1$  and  $u = 1, 2, \dots, U$ , which produce a modified data block  $X^u = [X^u|1, X^u|2, \dots, X^u|N-1]^T$ . IFFT of  $U$  independent sequences  $\{X^u|v\}$  are taken to produce the sequences  $x^u = [x^u|0, x^u|1, \dots, x^u|N-1]^T$ , among which the one  $\tilde{x} = x^{\tilde{u}}$  with the lowest PAPR is selected for transmission as shown as [6].

$$\tilde{u} = \underset{u=1,2,\dots,U}{\operatorname{arg\,min}} \left( \max_{n=0,1,\dots,N-1} |x^u|n| \right) \quad (2)$$

In order for the receiver to be able to recover the original data block, the information (index  $u$ ) about the selected phase sequence  $P^u$  should be transmitted as side information [9]. The implementation of SLM technique requires  $U$  IFFT operations. Furthermore, it requires  $\lceil \log_2 U \rceil$  bits of side information for each data block where  $\lceil x \rceil$  denotes the greatest integer less than  $x$ .

**IV. Simulation Result and Discussion**

It seems that the ability of PAPR reduction using SLM is affected by the phase sequence  $U$ , number of subcarriers  $N$  and  $N$ -point FFT/IFFT circuit where  $N$ -point FFT/IFFT circuit is required to the equal number of phase sequence  $U$ . Therefore, simulation with different values of  $U$  and  $N$  and the results exhibits some desired properties of signals representing the same information. The PAPR reduction performance of the proposed scheme is examined by computer simulation with MATLAB where over sampling factor is 4 and QAM mapping is adopted as modulation scheme in each sub-carrier. Phase sequences  $U=1, U=4, U=8$  and number of subcarriers  $N=32, N=64, N=128, N=256$  are used. OFDM is to split a high rate data stream into a number of lower rate data streams and modulated by subcarriers.

In Fig. 4, the  $N=32$  subcarriers are considering in this OFDM system. So 32 times bit duration increment of parallel data compare to the serial data stream. As a result inter symbol interference 32 times decreases. Hence data speed 32 times increase in this OFDM system. The complementary cumulative distribution function (CCDF) of PAPR of OFDM signals, where Selective Mapping (SLM) are used for PAPR reduction of OFDM signal and number of subcarrier 32. The

CCDF of the PAPR without SLM technique is approximately 12.2 dB where  $u=1$  means original PAPR of OFDM signal without SLM.

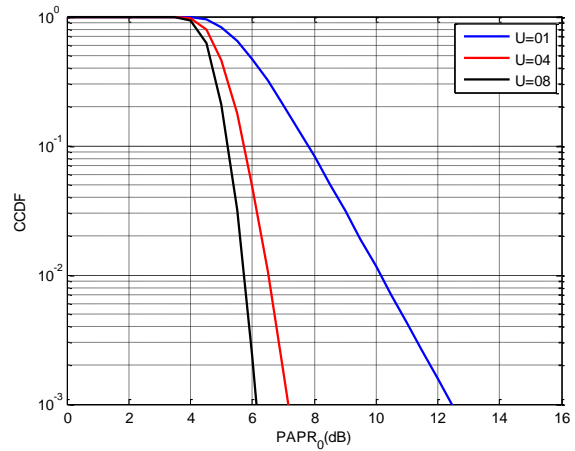


Fig. 4: PAPR performance for 32 sub carriers with AWGN channel

When  $u=4$  then PAPR of OFDM signal is 7.2 dB. The PAPR of OFDM signal is reduced by 5 dB. Again we consider  $u=8$  then PAPR of OFDM signal is 6.2 dB. So PAPR of OFDM signal more reduce by 6.2 dB. So it is clear that the probability of high PAPR is reduced significantly as the number of increase ( $U$ ).

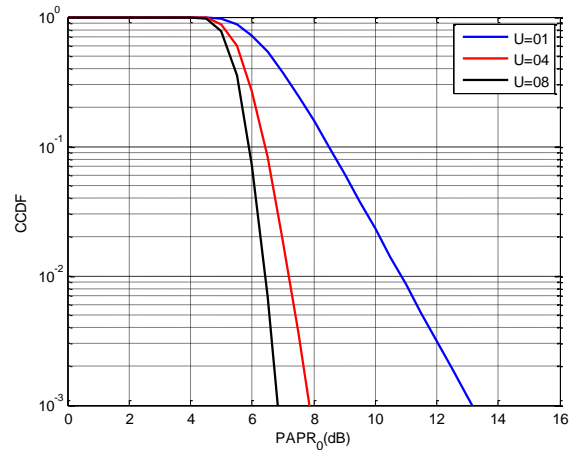


Fig. 5: PAPR performance for 64 sub carriers with AWGN channel

In Fig. 5, the  $N=64$  subcarriers are considering in this OFDM system. So 64 times bit duration increment of parallel data compare to the serial data stream. As a result inter symbol interference 64 times decreases. Hence data speed 64 times increase in this OFDM system. The complementary cumulative distribution function (CCDF) of PAPR of OFDM signals, where Selective Mapping (SLM) are used for PAPR reduction of OFDM signal and number of subcarrier 64. The CCDF of the PAPR without SLM technique is approximately 13.2 dB where  $u=1$  means original PAPR of OFDM signal without SLM. When  $u=4$  then PAPR of OFDM signal is 7.8 dB. The PAPR of OFDM

signal is reduced by 5.4 dB. Again we consider  $u=8$  then PAPR of OFDM signal is 6.6 dB. So PAPR of OFDM signal more reduce by 6.6 dB. So it is clear that the probability of high PAPR is reduced significantly as the number of increase ( $U$ ).

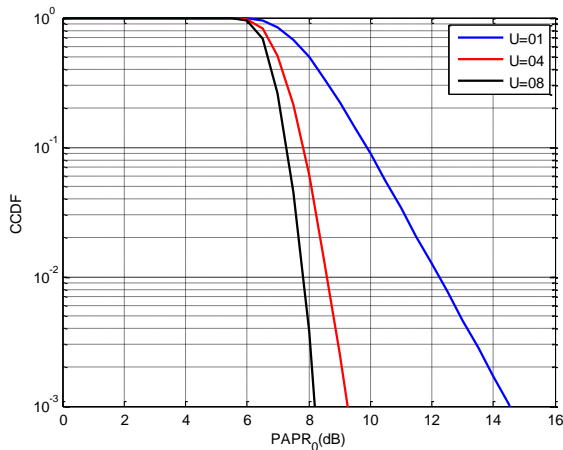


Fig. 6: PAPR performance for 128 sub carriers with AWGN channel

In Fig. 6, the  $N=128$  subcarriers are considering in this OFDM system. So 128 times bit duration increment of parallel data compare to the serial data stream. As a result inter symbol interference 128 times decreases. Hence data speed 128 times increase in this OFDM system. The complementary cumulative distribution function (CCDF) of PAPR of OFDM signals, where Selective Mapping (SLM) are used for PAPR reduction of OFDM signal and number of subcarrier 128. The CCDF of the PAPR without SLM technique is approximately 14.6 dB where  $u=1$  means original PAPR of OFDM signal without SLM. When  $u=4$  then PAPR of OFDM signal is 9.2 dB. The PAPR of OFDM signal is reduced by 5.4 dB. Again we consider  $u=8$  then PAPR of OFDM signal is 8.2 dB. So PAPR of OFDM signal more reduce by 6.4 dB. So it is clear that the probability of high PAPR is reduced significantly as the number of increase ( $U$ ).

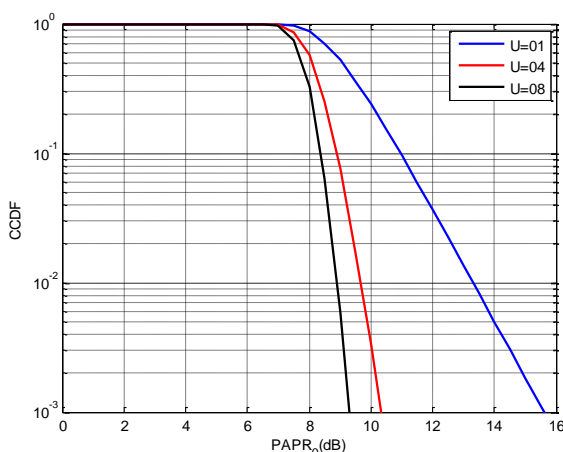


Fig. 7: PAPR performance for 256 sub carriers with AWGN channel

In Fig. 7, the  $N=256$  subcarriers are considering in this OFDM system. So 256 times bit duration increment of parallel data compare to the serial data stream. As a result inter symbol interference 256 times decreases. Hence data speed 256 times increase in this OFDM system. The complementary cumulative distribution function (CCDF) of PAPR of OFDM signals, where Selective Mapping (SLM) Technique are used for PAPR reduction of OFDM signal and number of subcarrier 256. The CCDF of the PAPR without SLM is approximately 15.6 dB where  $u=1$  means original PAPR of OFDM signal without SLM. When  $u=4$  then PAPR of OFDM signal is 10.2 dB. The PAPR of OFDM signal is reduced by 5.4 dB. Again we consider  $u=8$  then PAPR of OFDM signal is 9.2 dB. So PAPR of OFDM signal more reduce by 6.4 dB. So it is clear that the probability of high PAPR is reduced significantly as the number of increase ( $U$ ).

## V. Conclusion

In this paper, we studied the Performance of SLM for Different Number of Subcarriers. Simulation result shows that the PAPR is reduced significantly when the number of phase sequences is increased. We are clear from the above result, PAPR reduction is small difference for  $u=4$  with compared to  $u=8$  of selected mapping technique. When the phase sequence is increased then the number of FFT/IFFT circuit is increased in OFDM system. Implementation complexity and cost are increased. Hence  $u=4$  means four different phase sequence are conventionally taken for SLM to reduce the PAPR of OFDM signal. The  $N=32$ ,  $N=64$ ,  $N=128$  and  $N=256$  subcarriers are considering in this OFDM system respectively with different simulation results. So 32, 64, 128 and 256 times bit duration increment of parallel data compare to the serial data stream. As a result inter symbol interference 32, 64, 128 and 256 times decreases. Hence data speed 32, 64, 128 and 256 times increase in this OFDM system. So data speed depends on different kinds of subcarriers numbers. It shows that data speed increases when subcarriers increase. But  $N$ -point IFFT/FFT circuit depends on  $N$ -subcarriers. Complex multiplications in FFT algorithm  $(N/2)\log_2 N$  depends on the number of points  $N$ . For  $N=32$ ,  $N=64$ ,  $N=128$  and  $N=256$  numbers of point's complex multiplication are needed 80, 192, 448 and 1024 respectively. Hence when number of subcarriers  $N$  is increased, the complex multiplication in FFT is increased and when we use number of subcarriers  $N=128$  (Fig 5) where number of phase sequence  $u = 4$  the PAPR of OFDM signal is reduced by 5.4 dB and  $u=8$ , the PAPR of OFDM signal is reduced by 6.4 dB. Here 128-point IFFT/FFT is needed and complex multiplication is 448. So  $N=128$  are conventionally taken for SLM.

The selected technique provides us with a good range in performance to reduce PAPR. The degradation of

performance of the additional noise can be compensated by a smaller A/D-D/A quantization noise due to a reduced dynamical range through clipping. The proposed technique reduces PAPR, decrease the BER over conventional techniques, and improve the spectrum efficiency, where the proposed technique is evaluated in presence of nonlinear power amplifier. Although the some drawback is exist in the SLM. It has very high implementation complexity with high cost and side information is required. However the SLM technique is the best distortion less technique to reduce the high PAPR of OFDM signal. So high speed data transmission is possible if we used the SLM in that OFDM system.

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