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Design and Analysis of a Wearable Monopole Antenna on Jeans Substrate for RFID Applications

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Abstract

In this paper a novel rectangular C-shaped planar monopole antenna is designed and analyzed for Radio Frequency Identification (RFID) application in the licensed-free Industrial Scientific and Medical (ISM) band of 2.45 GHz. The proposed antenna uses a 1 mm thicker wearable Jeans fabric {relative permittivity, $\epsilon_r=1.68$ and Loss Tangent of $\delta=0.025$ } as a substrate material. For better bandwidth, and radiation efficiency the substrate is backed by a truncated copper ground plane. The antenna operate efficiently (92.85%), giving an adequate bandwidth, return-loss, gain and directivity of 12 %, -35.57 dB, 3.144 dB and 3.465 dBi, respectively. The nominated antenna is compact, low profile, and provides a better impedance matching (50.25 ohms) which results in a Voltage Standing Wave Ratio (VSWR) of 1.03. The far-field analysis is carried out via Finite Integration Technique (FIT) in CST Microwave studio.

Index Terms: Planar antenna, radio frequency identification, wearable, monopole, low profile.

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1. Introduction

With the advent in modern communication systems, RFID (Radio Frequency Identification) is graded at a higher place than Traditional Identification Methods (TIM), e. g. Bar Code Scanning, over the past decade leading a huge number of commercialization and research endeavor in the early 2000s [1]. Due to the excellent performance in the field of automatic identification and data collection, RFID technology is broadly used, and different frequency ranges have been assigned depending upon the RFID applications. The specific frequency ranges includes 120-150 KHz (LF-Unregulated), 13.56 MHz (HF-ISM BAND), 865-868 MHz (UHF-ISM BAND), (2.45 GHz) 2.400-2.483 GHz (Microwave-ISM BAND),(5.8 GHz) 5.725-5.875 GHz

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(Microwave-ISM BAND). The maximum data rate transfer and the higher ranges are achieved on higher frequencies (Microwave-Region). In RFID microwave (MW) region (2.45/5.8 GHz), the layout of an antenna becomes more knotty and crucial [2]. The Basic composition of the passive UHF and microwave frequencies (MWF), RFID systems are a tag with some stored information, linked to an object and the tag is activated by electromagnetic waves (EM) imparted by the reader. Communication is processed when the backscatter radio waves are modulated by the data stored in the tag. RFID tags are widely in various fields of automatic identification of objects such as access control, asset identification and industry manufacture and much more [3]. RFID is the most suitable way to communicate with wireless body area network (WBANs) through their backscattered radio waves. In back scattered system, the communication between the tag and the reader is ensured to be power efficient. The radio propagation channel and the efficiency of back scattered performance of on-body RFID system is obligatory to inspect in order to make sure, in reasonable ways the on-body RFID system's communication and the power delivered is durable [4].

The perception behind the idea of wearable computing systems become very trendy and it narrates the imminent of electronic systems as an elemental part of our clothes. The wearable systems, in the start, were used for navigation tasks, military operations, and surveillance purposes [5]. With the development in recent years, the technology has given us a podium in antenna that can be labelled as wearable antenna. These wearable antennas can be efficiently used for monitoring and indication. The ease of these labelled antennas is light weight, low cost and a reliable structure. The recent researches show that the flawless location for these antennas is the anatomy of the human body [6]. The life of human beings these days is glimmered with the sustained monitoring system. This technology is simply the imminent of smart garments and human life [7]. Due to the appealing characteristics (reliable structure, low-profile, light weight) of these antennas, researchers aim is to design an efficient wearable monopole antenna [8-9]. Radiating elements of various shapes have been designed till time e-g E-shaped [10], B-shaped [11], F-shaped [12], G-shaped [13], Double-T-shaped [14] and triangular shape [15]. In [16] a compact RFID antenna, operating in the unlicensed 60 GHz millimeter wave band is presented. The advantage of the 60 GHz frequency is the availability of the higher bandwidth in this band, which can support higher data rates and more RFID channels and services. The disadvantage of this technology is that the size of the antenna is extremely small at very high frequency (60 GHz) which makes the fabrication quite challenging.

It is evident from Fig. 1 that the RFID tag antenna technology is an important component of modern industrial sectors [1] mainly used in:

- Manufacturing, logistics, distribution
- Asset management
- Access and quality control
- Payment
- Product, inventory tracking and item location
- Identification and inspection

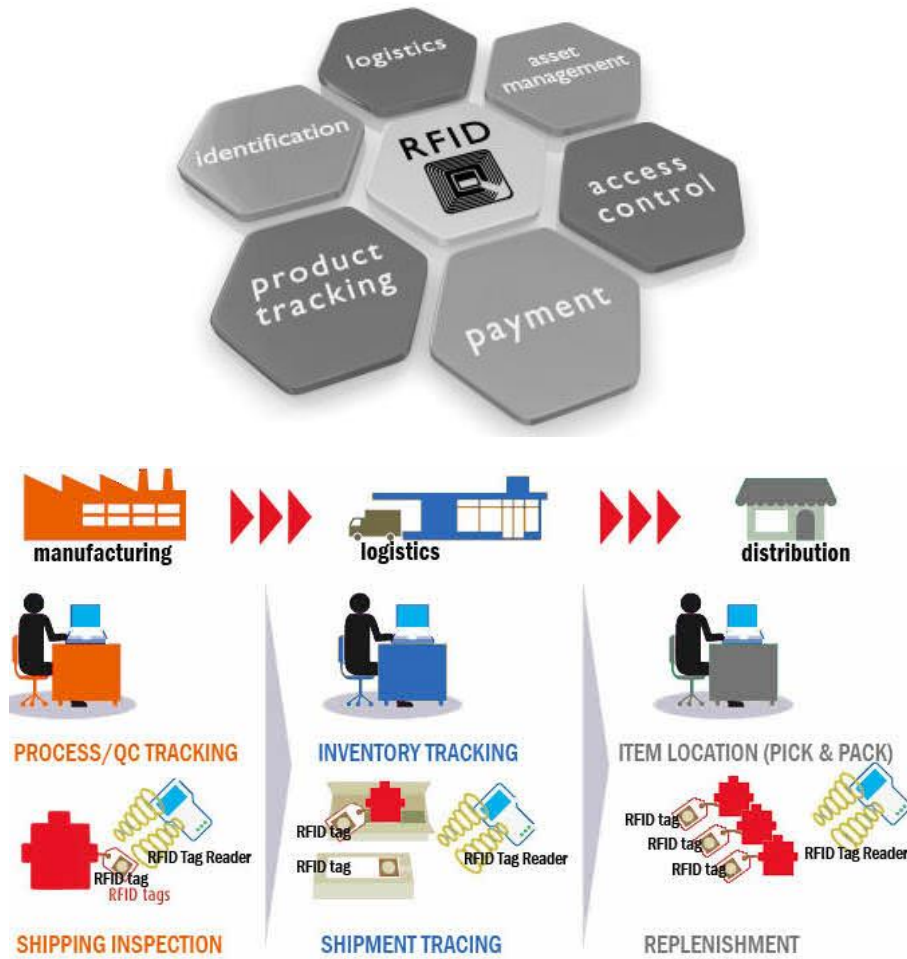


Fig.1. Applications of RFID Tag antennas

In this paper a rectangular C-shaped antenna is designed on 1-mm thick Jeans substrate with a truncated copper ground plane. The main objective is to design and analyze a 2.45 GHz wearable monopole planar antenna for RFID applications. On the basis of simulated results, it is found that the nominated antenna gives superior performance in context of efficiency, reflection coefficient (S_{11}), directivity, gain and bandwidth.

The rest of the research work is organized as follow: Segment 2 shows the theory and layout of planar monopole antenna. Segment 3 portrays the simulation and calculated results. The conclusion and future work of this research is stated in segment 4.

2. Antenna Layout and Theory

This section explains the layout and design theory of the proposed RFID antenna. The antenna is designed using the standard transmission line theory [17]. The antenna is excited using a waveguide port assigned to a microstrip feed line of width (FWI). For better bandwidth and efficiency, a truncated metallic ground plane is used beneath the substrate.

2.1. Layout

The layout of the rectangular C-shaped, 2.45 GHz RFID antenna is demonstrated in Fig. 2. The proposed antenna is backed by a 1 mm thicker Jeans substrate. A trimmed copper ground plane is used in order to achieve optimal efficiency and gain. The feed of the antenna is a 50Ω copper micro-strip line. The width (FW=5 mm) of the microstrip feed-line is calculated using [17]. The feed-line is connected with the radiating element of the antenna at a convenient driving point position, which results in a good impedance matching, low reflection coefficient and voltage standing wave ratio. In other words for better radiation efficiency the driving point impedance of the antenna must be closely comparable with the feed-line impedance.

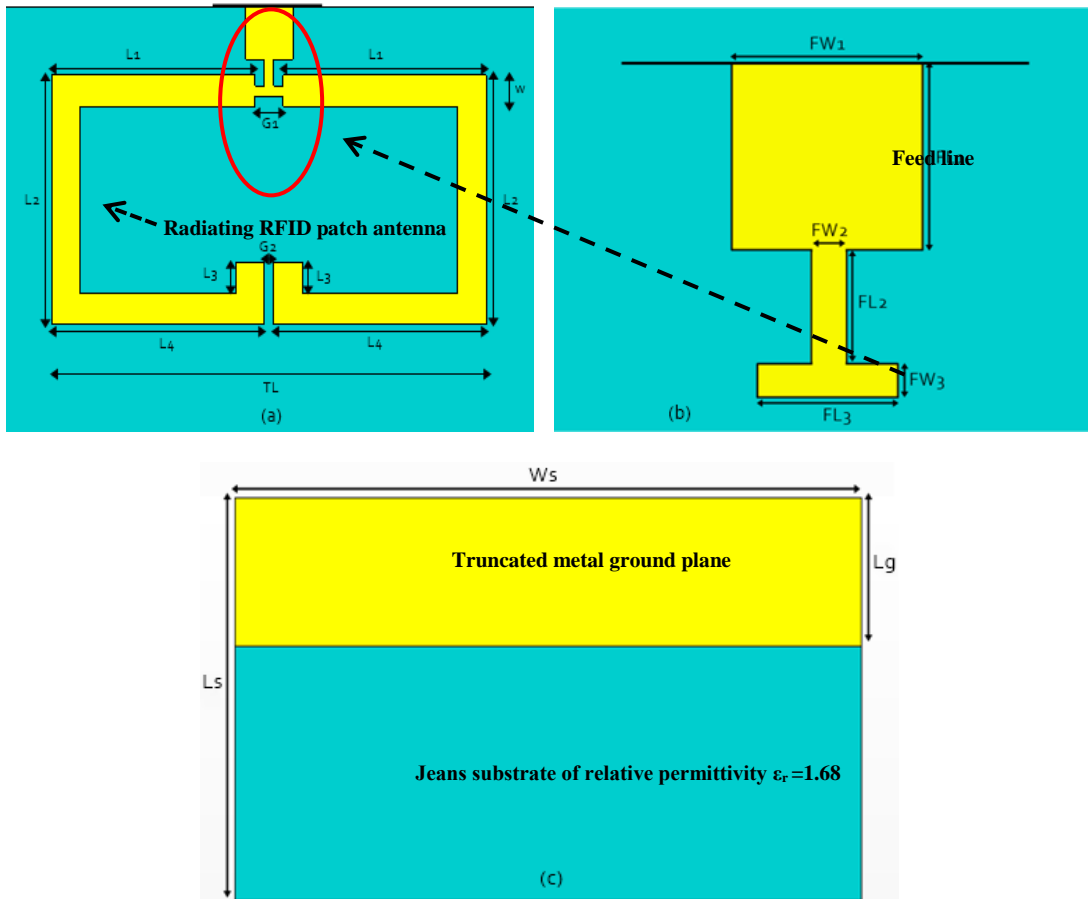


Fig.2. Layout of the proposed RFID antenna (a) front view (b) Zoomed View of Feed, (c) Rear View

In addition to compact size, the antenna is designed using a flexible and wearable jeans substrate material, which makes it a suitable candidate for body worn RFID applications.

The various dimensions of the proposed antenna are encapsulated in Table 1.

Table 1. The dimensional parameters of Fig. 2

Parameters	Values (mm)	Parameters	Values (mm)
L1	18.5	FL1	10
L2	24	FL2	5
L3	2.3	FL3	3
L4	19.5	FW1	5
G1	3	FW2	1
G2	1	FW3	1
W	3	TL	40
Ws	90	Lg	18
Ls	50	hs	1

2.1. Theory

Recalling the theory of transmission lines [17], effective length of the nominated antenna is found, that is $L_{2.45} = 140.6 \text{ mm} = 2(L1+L2+L3+L4+2W)$. The resonant length, $L_{2.45}$, in terms of guided wavelength, $\lambda_{2.45}$ is found using the following equation:

$$L_{2.45} = \frac{\lambda_{2.45}}{4} \quad (1)$$

Where, $\lambda_{2.45}$ is the guided wavelength of the waves within the substrate at resonant frequency, $f_r=2.45 \text{ GHz}$. It is found using:

$$\lambda_{2.45} = \frac{c}{f_r \sqrt{\epsilon_r}} \quad (2)$$

Where, c is speed of light, ϵ_r is effective permittivity of the substrate [17]. The inclusive volume of the nominated antenna is $90 * 50 * 1 \text{ mm}^3$.

The dimensions of the proposed antenna are optimized for better radiation efficiency (η_{rad}), which is defined as the ratio of the radiated power of the antenna to its input power.

$$\eta_{rad} = \frac{\text{Radiated power}}{\text{Input Power}} \quad (3)$$

The efficiency is higher if the antenna is fed at the right point, resulting in minimum value of reflection coefficient (Γ). This factor is the ratio of reflected and incident electric fields. Mathematically it is quantified as:

$$|\Gamma| = \frac{Z_a - Z_0}{Z_a + Z_0} \quad (4)$$

Where, Z_a is the driving point impedance of the antenna and Z_0 is the characteristic impedance of the microstrip feed line. Reflection coefficient, voltage standing wave ratio (VSWR) and return loss are interrelated parameters of the antenna, i.e. mathematically:

$$VSWR = \frac{1+|\Gamma|}{1-|\Gamma|} \quad (5)$$

If reflection coefficient is small then VSWR approaches unity. The return loss is the ratio of the reflected power to the incident power of the antenna and is usually expressed in decibel (dB) scale.

$$\text{Return Loss (dB)} = \frac{\text{Reflected Power}}{\text{Input Power}} = -20 \log_{10} |\Gamma| \quad (6)$$

Lower value of return loss at the design frequency is achieved if the antenna is properly matched (i.e. low Γ). The gain (G) and directivity (D) of the proposed antenna are related by the radiation efficiency as:

$$G = \eta_{rad} D \quad (7)$$

3. Results

The nominated C-Shaped monopole antenna is sketched and inspected by a technique called FIT (Finite Integration Technique) in CST Microwave Studio [18]. The boundary conditions are taken as ‘open add space’ and using normal (free-space) background material. The transient solver is used to analyze the antenna performance parameters (i.e. return loss, VSWR, gain, directivity and surface electric field plots).

The antenna is excited through a waveguide port assigned to the 5 mm wide transmission feed line. Scattering parameters, return-loss, surface electric fields, gain and directivity in E and H-planes are figured out using open add space boundary conditions using the transient solver. The nominated antenna operates at 2.45 GHz with a return loss (RL) of -35.57 dB. The antenna gives bandwidth of 12% at 2.45 GHz, relative to return loss ≤ -10 dB, as highlighted in Fig. 3.

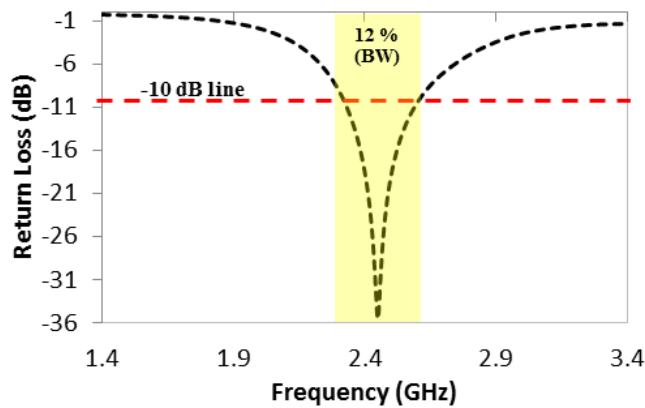


Fig.3. S_{11} Return loss(RL) in single frequency mode.

The Voltage Standing Wave Ratio (VSWR) of the nominated antenna is 1.03 at 2.45 GHz (Fig. 4). It shows that the nominated antenna is perfectly matched i.e. having driving point impedance of 50.25Ω at 2.45 GHz frequency. Hence the available input power to the antenna is predominantly radiated in the desired direction, with no (or extremely low power) reflected in the reverse direction. In other words the propagation of the standing waves along the feed line is highly restricted due to the impedance matching which helps in achieving very good radiation efficiency by the proposed RFID antenna.

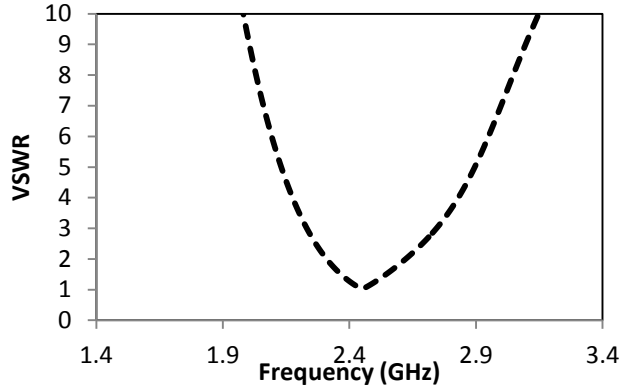


Fig.4. VSWR of the proposed antenna

The directivity patterns of the nominated monopole antenna in the E and H-plane at 2.45 GHz are portrayed in Fig. 5a. The nominated antenna gives a peak value of E-plane directivity (3.465 dBi) at 2.45 GHz. The gain patterns of the nominated monopole antenna in the E and H-plane at 2.45 GHz is depicted in Fig. 5b. The antenna gives an E-plane gain of 3.144 dB at 2.45 GHz. The antenna radiates maximum power along the boresight direction ($\theta=0^\circ$). These plots show that the given antenna radiates predominantly Omni-directionally in the H-plane. A single null in radiation pattern appears at 90° in the E-plane radiation pattern of the antenna. The directivity and gain of the proposed antenna can be further enhanced by using:

- An artificial metamaterial ground plane as a replacement to the truncated ground plane
- Replacing the single radiating element by a two or four element array of identical elements (i.e. array antenna)

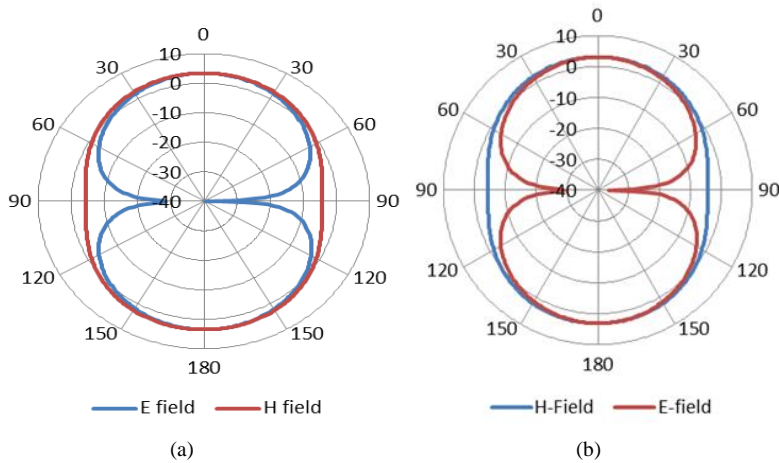


Fig.5. E and H-planes pattern of the proposed antenna at 2.45 GHz (a) Directivity (b) Gain

The radiated electric (E) and magnetic (H) fields of the antenna at 2.45 GHz are portrayed in Fig. 6a and b respectively. These plots clearly show the radiating length of the nominated antenna. The E field is higher in the center of the lower (horizontal) arm, while minimum along the side (vertical) arms. The H field is inversely related to the E field therefore its intensity is higher along the vertical arms of the proposed antenna.

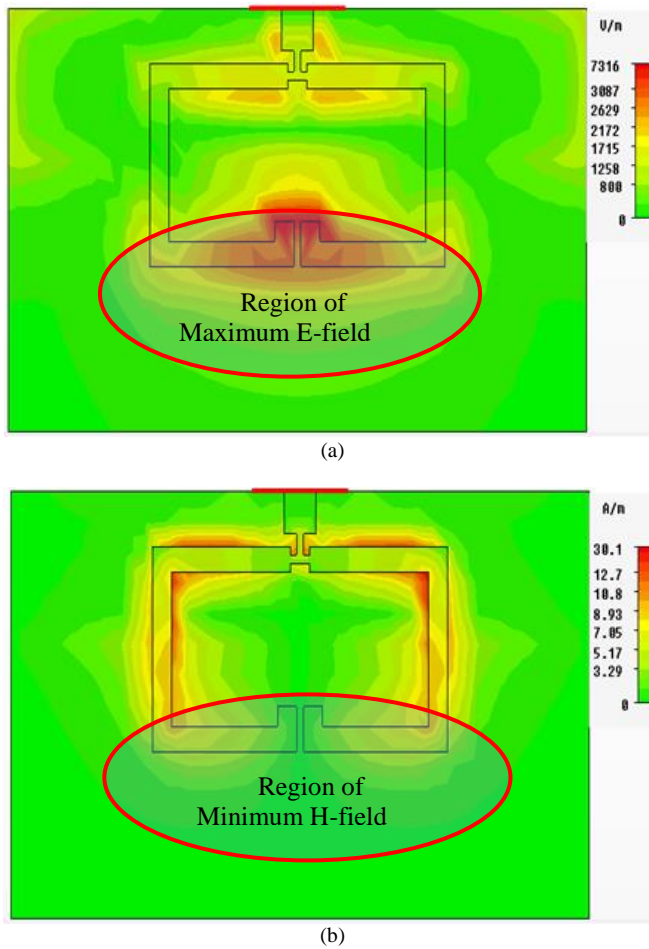


Fig.6. Radiated (a) E field and (b) H field, at 2.45 GHz

The three-dimensional plots of directivity and gain of the nominated antenna are portrayed in Fig. 7 and 8 respectively. A peak directivity and gain of the RFID antenna are 3.42 dBi and 3.1 dB respectively at 2.45 GHz. The nominated antenna has an extremely good efficiency of 92.8% at 2.45 GHz (Centered Frequency), due to the matched design procedure. The summary of the results is portrayed in Table 2 below:

Table 2. Summed up results

Parameters	Results
Frequency (GHz)	2.45
Directivity (dBi)	3.42
Gain (dB)	3.144
Return Loss (dB)	-35.57
Bandwidth (%)	12
Efficiency (%)	92.8
Impedance (Ω)	50.25
VSWR (ratio)	1.03

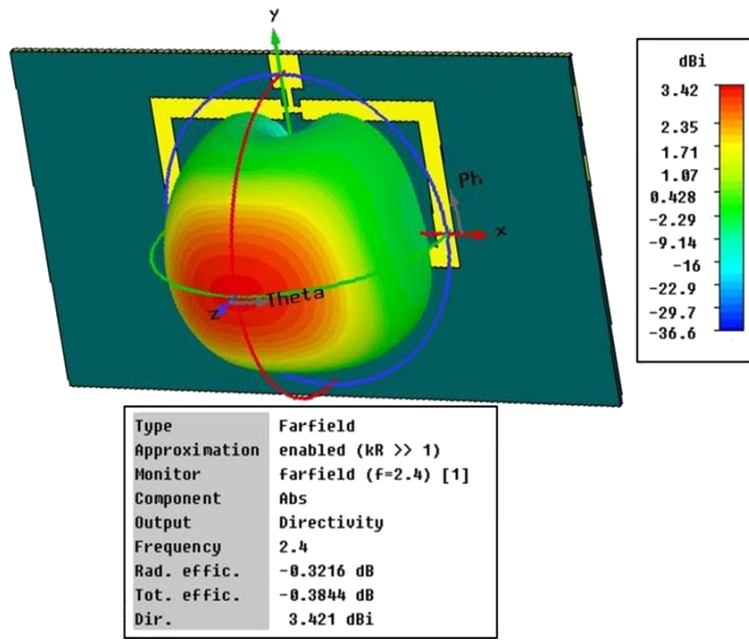


Fig.7. Far-field directivity pattern (3D) at 2.45 GHz

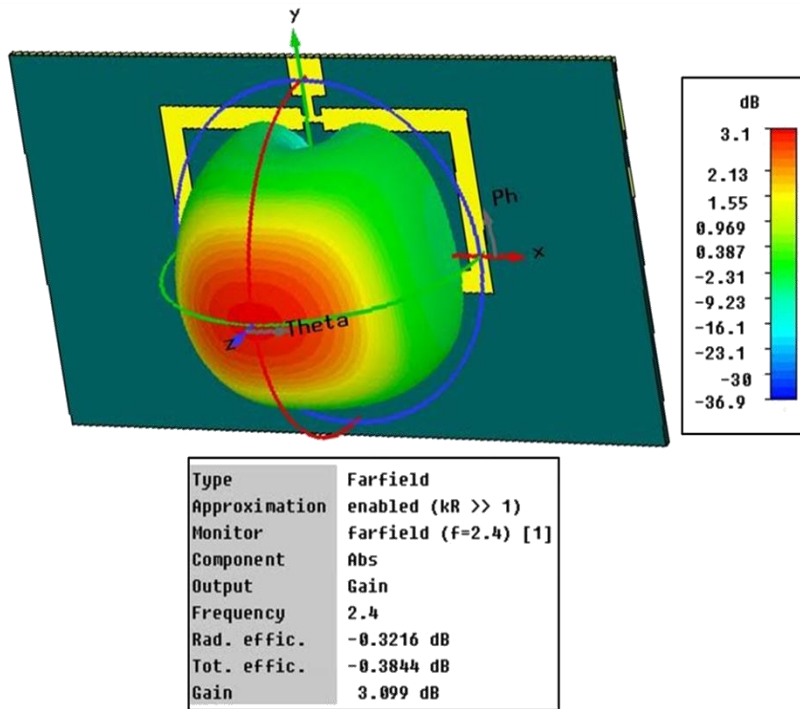


Fig.8. Far-field gain pattern (3D) at 2.45 GHz

The three dimensional far-field plots clearly show that the antenna radiates most of the power orthogonal to the plane of the antenna geometry. Negligible amount of power is radiated in the backward direction. Therefore the antenna can be safely used in wearable applications which demands for low level of backward radiations towards human body.

4. Conclusions

A single-band wearable and planar rectangular C-shaped monopole antenna for RFID (2.45 GHz) has been presented in this paper. The C-shaped planar monopole antenna work highly efficiently (92.8 %) at 2.45 GHz, due to a truncated copper ground plane and extremely matched feeding mechanism.

Bandwidth of 12% (294 MHz), relative to -10 dB threshold of return loss, has been achieved at 2.45 GHz. The far-field pattern of the nominated antenna is acclaimed to be omni-directional, with a relatively good gain of 3.144 dB due to good impedance matching at 2.45 GHz. The nominated C-shaped monopole antenna can be used in RFID applications such as medical implants, remote patient monitoring, rescue and military operations.

The truncated ground plane can be replaced by an electromagnetic bandgap structure for further better gain, directivity, efficiency and reduced level of Specific Absorption Rate (SAR) in wearable scenarios. The effects of bending and crumpling on antenna performance parameters can also be studied in body worn conditions. Furthermore the antenna can be reconfigured in dual and triple band modes, using optical or PIN diode switches in order to add other dedicated RFID frequency bands. The prototype of the proposed antenna will be fabricated to validate the numerical results.

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